

A.

1. The transfer function is

$$T(s) = \frac{KK_m}{s + 1/\tau_m} \cdot \frac{1}{1 + \frac{KK_m}{s + 1/\tau_m}} = \frac{KK_m}{s + \frac{1}{\tau_m} + KK_m}$$

Pole is in the left hand so we can use the final value theorem.

$$\mathcal{U}_0(s) = \mathcal{U}_i(s) \cdot T(s) = \frac{1}{s} \frac{KK_m}{s + \frac{1}{\tau_m} + KK_m}$$

Final value theorem:

$$\omega_o(\infty) = \lim_{s \rightarrow 0} s \mathcal{U}_0(s) = \frac{KK_m}{\frac{1}{\tau_m} + KK_m} < 1$$

$$\text{want } \frac{|\omega_o(\infty) - \omega_i|}{\omega_i} = 0.05$$

$$\omega_i = 1 \Rightarrow 1 - \frac{KK_m}{\frac{1}{\tau_m} + KK_m} = 0.05$$

$$\text{or } \frac{KK_m}{\frac{1}{\tau_m} + KK_m} = 0.95$$

$$\Rightarrow \frac{1}{\frac{1}{\tau_m KK_m} + 1} = 0.95$$

$$\Rightarrow 1 = 0.95 \left(\frac{1}{\tau_m KK_m} + 1 \right)$$

$$\Rightarrow \frac{1}{\tau_m KK_m} = \frac{0.05}{0.95} = 0.053$$

$$\Rightarrow \tau_m K K_m = 19$$

$$\therefore K = \frac{19}{\tau_m K_m} = \frac{19}{150 \cdot 0.4} = 0.32$$

Motor speed as a function of time:

$$\Omega_o(s) = \frac{K K_m}{s(s+a)}$$

$$a = \frac{1}{\tau_m} + K \cdot K_m$$

Partial fractions

$$\Omega_o = \frac{A}{s} + \frac{B}{s+a}$$

$$\Rightarrow A = \frac{K K_m}{a}$$

$$B = \frac{K K_m}{-a}$$

$$\Rightarrow \omega_o(t) = \frac{K K_m}{a} (1 - e^{-at})$$

$$\text{where } a = \frac{1}{0.4} + 150 \cdot (0.32) = 50, K K_m = 47.5$$

$$\omega_o(t) = 0.95 (1 - e^{-50t})$$

2. The transfer function is the same as in part 1 except we need to replace K with $K/s \Rightarrow$

$$T(s) = \frac{\frac{K}{s} K_m}{(s + \frac{1}{\tau_m} + \frac{K}{s} K_m)} = \frac{K K_m}{s^2 + \frac{s}{\tau_m} + K K_m}$$

From class we know that the damping ratio and natural frequency are

$$\omega_0 = \sqrt{k \cdot K_m} \quad (1)$$

$$\zeta = \frac{1/\tau_m}{2 \omega_0} \quad (2)$$

Also the fractional overshoot is given by the equation

$$OS = e^{-\frac{\zeta \pi}{\sqrt{1-\zeta^2}}} \quad (3)$$

Thus, we want $e^{-\zeta \pi / \sqrt{1-\zeta^2}} = 0.05$

$$\Rightarrow \frac{\zeta}{\sqrt{1-\zeta^2}} = \frac{3}{\pi} = 0.95$$

$$\Rightarrow \frac{\zeta^2}{1-\zeta^2} = 0.91 \Rightarrow \zeta^2 = .91 - .91 \zeta^2$$

$$\Rightarrow 1.91 \zeta^2 = .91 \Rightarrow \zeta = 0.69 \quad (4)$$

$$\therefore (1), (2), (4) \Rightarrow \frac{1/\tau_m}{2 \sqrt{k \cdot K_m}} = 0.69$$

$$\Rightarrow \sqrt{k K_m} = \frac{1}{\tau_m \cdot 2 \cdot 0.69} = 1.81$$

$$\Rightarrow K = \frac{3.28}{K_m} = \frac{3.28}{150} = 0.0219$$

Dynamic Response:

$$\begin{aligned}\text{Then } T(s) &= \frac{(0.0219)(150)}{s^2 + 2.5s + (0.0219)(150)} \\ &= \frac{3.28}{s^2 + 2.5s + 3.28}\end{aligned}$$

Laplace transform of output

$$\Theta_o(s) = \frac{1}{s} T(s) = \frac{1}{s} \cdot \frac{3.28}{s^2 + 2.5s + 3.28}$$

$$\text{Let } \Theta_o(s) = \frac{A}{s} + F(s)$$

$$A = 1 \Rightarrow F(s) = \frac{3.28 - (s^2 + 2.5s + 3.28)}{s(s^2 + 2.5s + 3.28)}$$

$$= - \frac{s + 2.5}{s^2 + 2.5s + 3.28}$$

Complete the square for the denominator.

$$s^2 + 2.5s = (s + 1.25)^2 - (1.25)^2$$

$$3.28 - 1.25^2 = 1.72$$

$$\sqrt{1.72} = 1.31 \equiv \omega$$

$$\therefore F(s) = - \frac{s + 2.5}{(s + 1.25)^2 + \omega^2}$$

The Laplace transform tables in book $\Rightarrow$

(Note $B = -1$, $C = -2.5$, $\omega = 1.31$, $a = 1.25$)

$$f(t) = - \left[\cos \omega t + \left(\frac{2.5 - 1.25}{1.31} \right) \sin \omega t \right] e^{-1.25t}$$

Dynamic Response

$$\theta_0(t) = 1 - e^{-1.25t} \left(\cos \omega t + 0.95 \sin \omega t \right)$$

$$\omega = 1.31$$

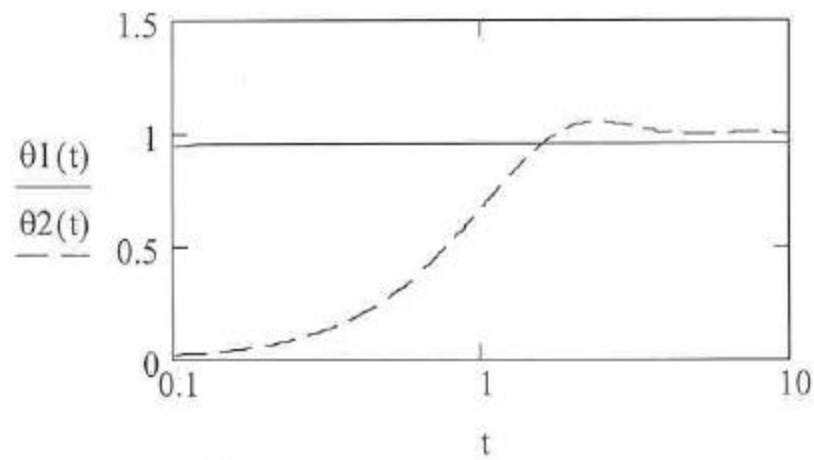
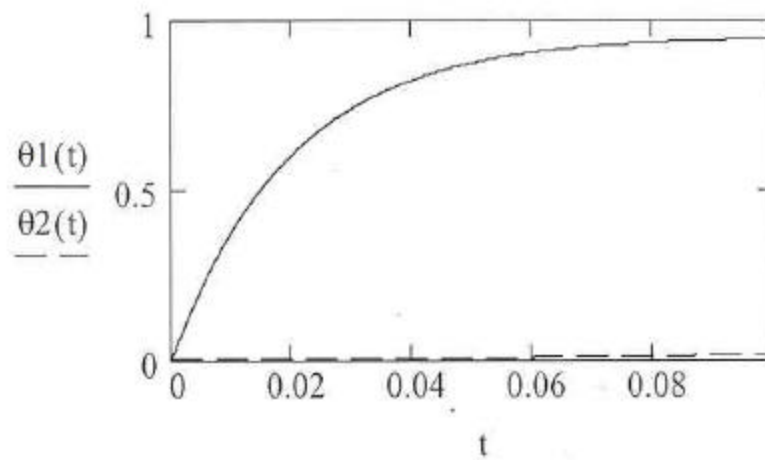
See plots from Mathcad on next page.

Proportional Control

$$\theta_1(t) := 0.95 \cdot (1 - e^{-50 \cdot t})$$

Integral Control $\omega := 1.31$

$$\theta_2(t) := 1 - e^{-1.25 \cdot t} \cdot (\cos(\omega \cdot t) + 0.95 \cdot \sin(\omega \cdot t))$$



B.

• Equations of motion

$$m \ddot{x}_1 = -(1)(x_1 - x_2) - (1) \cdot x_1 = -2x_1 + x_2$$

$$m \ddot{x}_2 = -(1)(x_2 - x_1) = -x_2 + x_1$$

In matrix form

$$m \ddot{\vec{x}} + K \vec{x} = \vec{0}$$

$$K = \begin{bmatrix} 2 & -1 \\ -1 & 1 \end{bmatrix}$$

Assume $\vec{x} = \vec{b} e^{j\omega t}$ (or $\vec{x} = \vec{b} \sin \omega t$ or $\vec{b} \cos \omega t$)

$$\ddot{\vec{x}} = -\omega^2 \vec{b} e^{j\omega t}$$

$$\therefore -m\omega^2 \vec{b} e^{j\omega t} + K \vec{b} e^{j\omega t} = \vec{0}$$

$$\Rightarrow K \vec{b} = m\omega^2 \vec{b}$$

$$\text{or } [K - m\omega^2 I] \vec{b} = \vec{0}$$

Either $\vec{b} = \vec{0}$ or $\det [K - m\omega^2 I] = 0$ $m=1$

$$\Rightarrow \begin{vmatrix} (2 - \omega^2) & -1 \\ -1 & (1 - \omega^2) \end{vmatrix} = 0 \Rightarrow (2 - \omega^2)(1 - \omega^2) - 1 = 0$$

This implies $\omega^4 - 3\omega^2 + 2 - 1 = 0$
or $\omega^4 - 3\omega^2 + 1 = 0$

$$\text{Quadratic formula} \Rightarrow \omega^2 = \frac{3 \pm \sqrt{9 - 4 \cdot 1}}{2} = \frac{3 \pm \sqrt{5}}{2}$$

Mode shape for lower frequency

$$\omega^2 = \omega_1^2 = \frac{3-\sqrt{5}}{2} = 0.382 \Rightarrow \omega_1 = 0.618$$

$$(2 - \omega_1^2) b_1 - b_2 = 0$$

$$\Rightarrow \left(2 - \frac{3-\sqrt{5}}{2}\right) b_1 = b_2$$

$$\Rightarrow (2 - 0.382) b_1 = b_2 \Rightarrow b_2 = 1.618$$

$$\therefore \vec{b}_1 = c_1 \begin{bmatrix} 1 \\ 1.618 \end{bmatrix} \quad (\text{Note: any multiple of this vector is correct})$$

Mode shape for higher frequency; $\omega^2 = \omega_2^2 = \frac{3+\sqrt{5}}{2} = 2.618$

$$\omega_2 = 1.618$$

$$(2 - \omega_2^2) b_1 - b_2 = 0$$

$$(2 - 2.618) b_1 = b_2 \Rightarrow b_2 = -0.618 b_1$$

$$\vec{b}_2 = c_2 \begin{bmatrix} 1 \\ -0.618 \end{bmatrix} \quad (\text{Note: any multiple of this vector is correct.})$$